



INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(51) International Patent Classification 6 : H04B 1/707	A2	(11) International Publication Number: WO 99/38270
		(43) International Publication Date: 29 July 1999 (29.07.99)

(21) International Application Number: PCT/GB99/00130

(22) International Filing Date: 15 January 1999 (15.01.99)

(30) Priority Data:

9801386.5	22 January 1998 (22.01.98)	GB
98300440.9	22 January 1998 (22.01.98)	EP

(71) Applicant (for all designated States except US): BRITISH TELECOMMUNICATIONS PUBLIC LIMITED COMPANY [GB/GB]; 81 Newgate Street, London EC1A 7AJ (GB).

(72) Inventor; and

(75) Inventor/Applicant (for US only): WIDDOWSON, Terence [GB/GB]; 37 Sproughton Court, Sproughton, Ipswich IP8 3AJ (GB).

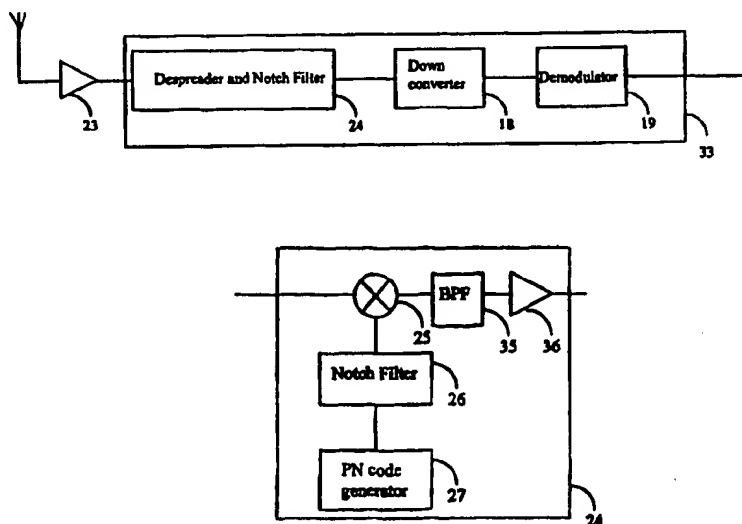
(74) Agent: SEMOS, Robert, Ernest, Vickers; BT Group Legal Services, Intellectual Property Dept., Holborn Centre, 8th floor, 120 Holborn, London EC1N 2TE (GB).

(81) Designated States: CN, IN, JP, KR, SG, US, European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE).

Published

Without international search report and to be republished upon receipt of that report.

(54) Title: RECEIVING SPREAD SPECTRUM SIGNALS WITH NARROWBAND INTERFERENCE



(57) Abstract

A receiver receives spread spectrum signals which also include unwanted narrowband signals, for example, radio signals which comprise a mixture of signals from signal sources using the Group Special Mobile (GSM) ETSI standard coding and from signal sources using Code Division Multiple Access (CDMA) coding. In the invention both despreading of the spread spectrum signal and attenuating of predetermined frequencies (corresponding to the narrowband signal) of the spread spectrum signal are performed prior to downconverting. In another approach to decoding a spread spectrum signal in which an unwanted narrowband signal has been coded using a coding scheme with error correction capabilities the narrowband signal is decoded, regenerated and subtracted prior to despreading of the spread spectrum signal.

FOR THE PURPOSES OF INFORMATION ONLY

Codes used to identify States party to the PCT on the front pages of pamphlets publishing international applications under the PCT.

AL	Albania	ES	Spain	LS	Lesotho	SI	Slovenia
AM	Armenia	FI	Finland	LT	Lithuania	SK	Slovakia
AT	Austria	FR	France	LU	Luxembourg	SN	Senegal
AU	Australia	GA	Gabon	LV	Latvia	SZ	Swaziland
AZ	Azerbaijan	GB	United Kingdom	MC	Monaco	TD	Chad
BA	Bosnia and Herzegovina	GE	Georgia	MD	Republic of Moldova	TG	Togo
BB	Barbados	GH	Ghana	MG	Madagascar	TJ	Tajikistan
BE	Belgium	GN	Guinea	MK	The former Yugoslav Republic of Macedonia	TM	Turkmenistan
BF	Burkina Faso	GR	Greece			TR	Turkey
BG	Bulgaria	HU	Hungary	ML	Mali	TT	Trinidad and Tobago
BJ	Benin	IE	Ireland	MN	Mongolia	UA	Ukraine
BR	Brazil	IL	Israel	MR	Mauritania	UG	Uganda
BY	Belarus	IS	Iceland	MW	Malawi	US	United States of America
CA	Canada	IT	Italy	MX	Mexico	UZ	Uzbekistan
CF	Central African Republic	JP	Japan	NE	Niger	VN	Viet Nam
CG	Congo	KE	Kenya	NL	Netherlands	YU	Yugoslavia
CH	Switzerland	KG	Kyrgyzstan	NO	Norway	ZW	Zimbabwe
CI	Côte d'Ivoire	KP	Democratic People's Republic of Korea	NZ	New Zealand		
CM	Cameroon	KR	Republic of Korea	PL	Poland		
CN	China	KZ	Kazakhstan	PT	Portugal		
CU	Cuba	LC	Saint Lucia	RO	Romania		
CZ	Czech Republic	LJ	Liechtenstein	RU	Russian Federation		
DE	Germany	LK	Sri Lanka	SD	Sudan		
DK	Denmark	LR	Liberia	SE	Sweden		
EE	Estonia			SG	Singapore		

RECEIVING SPREAD SPECTRUM SIGNALS WITH NARROWBAND INTERFERENCE

This invention relates to receivers for receiving signals which comprise a mixture of signals from a plurality of sources where two of the signal sources use different coding schemes from each other. In particular this invention relates to receivers for receiving spread spectrum signals which also include unwanted narrowband signals, for example, radio signals which comprise a mixture of signals from signal sources using the Group Special Mobile (GSM) ETSI standard coding and from signal sources using Code Division Multiple Access (CDMA) coding.

10

Demand for systems and services which require the use of the radio spectrum has recently far outstripped the capacity of present systems. This means that future systems must provide significantly improved spectral efficiency. One technique which has been suggested is to allow wideband spread spectrum signals to share a common spectrum with conventional narrowband signals.

Spread spectrum signals use a bandwidth which is in excess of the minimum bandwidth required to transmit the information in the signal. One method of generating a spread spectrum signal is known as pseudo random noise modulation (or direct sequence modulation). The spectrum of a signal is spread using a spreading code at the transmitter. A corresponding receiver uses the same code to despread the spectrum. The technique used to despread the spectrum to regenerate the original data signal has the effect of spreading the spectrum of any other signals which are not correlated with the spreading code used by the transmitter. Where, as is usual, several wideband signals from different sources (using different codes) share the same frequency spectrum, this technique is known as Code Division Multiple Access (CDMA).

The transmitted narrowband signal generally has a high power spectral density compared with that of the transmitted wideband signal. In the narrowband signal receiver the wideband signal creates interference in the narrowband signal. Equally, once the combined received signal has been decoded by a spread spectrum receiver, the narrowband signal causes interference in the decoded spread spectrum signal.

One method which has been used to reduce the interference between the narrowband signal and the wideband signal is to use corresponding filters in the spread spectrum transmitter and receiver to attenuate those frequencies used by the narrowband signal. A notch filter in the spread spectrum transmitter reduces the interference in the narrowband signal from the wideband signal. A corresponding notch filter in the spread spectrum receiver attenuates the frequencies used by the narrowband signal prior to despreading the spectrum to decode the spread spectrum signal.

10

There are problems with using such techniques to reduce the cross signal interference. To minimise the signal removed by the attenuation filters which is outside the band used by the narrowband signal, filters with very sharp cut-offs are required.

15

However, in a system such as the GSM cellular network adjacent cells have transmitters using different frequencies. Therefore the receiver is required to receive different frequency bands according to which transmitter is being used. Filters which are required to attenuate these narrowband frequencies need to be programmable. Programmable filters are most easily achieved with digital hardware which implies the use of high order finite impulse response (FIR) filters. Such filters currently only work on signals up to a maximum of a few tens of megahertz so it is most convenient to implement such filtering at baseband.

20

25 Milstein and Schilling. "The CDMA Overlay Concept", International Symposium on Spread Spectrum Technologies and Applications 1996, disclose a spread spectrum transmitter and receiver which utilise a notch filter. The receiver performs filtering simultaneously with despreading at baseband.

30 Any non-linear distortion to the signal prior to attenuating those frequencies used by the narrowband signal means that any signal in the narrowband frequency range introduces noise outside that range. The power spectral density of the narrowband signal is typically many tens of dB higher than the power spectral density of the wideband signal. Any components between the radio frequency

receiver and the filter must have a very large dynamic range in order to minimise noise introduced into the remaining wideband signal by such non-linear distortions.

If a non-linear coding method is used to increase the dynamic range of such components then there are resolution problems for the low-level wideband signal.

Therefore the problem with known systems which utilise an attenuating filter, is that performing filtering at baseband requires components prior to the filtering step to have a very large dynamic range, but filtering at high frequency is not possible using currently available digital programmable filters.

According to the present invention there is provided an apparatus for decoding a spread spectrum signal comprising

means for generating a despreading signal for despreading the spectrum of a signal;

means for despreading the spectrum of a signal using the despreading signal;

means for downconverting a signal to produce a lower frequency signal; and

means for attenuating predetermined frequencies of a signal;

wherein the despreading means is connected to despread the spread spectrum signal prior to downconverting and the attenuating means is connected to attenuate predetermined frequencies of the spread spectrum signal prior to downconverting.

Preferably the received signal is a composite signal comprising a spread spectrum signal component and a narrowband signal component which has a much higher power spectral density than the spread spectrum signal component and the attenuated frequencies are those present in the narrowband signal component. Optionally the spectrum is de-spread using a pseudo noise code.

Advantageously, the attenuating means is connected to attenuate predetermined frequencies of the despreading signal and the resulting notched despreading signal

is connected to simultaneously despread the spread spectrum signal and attenuate the predetermined frequencies of the spread spectrum signal.

According to another aspect of the invention there is provided an apparatus for
5 decoding a spread spectrum signal comprising means for generating a desreading
signal for desreading the spectrum of the signal; and attenuating means
connected to attenuate predetermined frequencies of the desreading signal to
provide a notched desreading signal; characterised in that the apparatus further
comprises mixing means connected to mix the notched desreading signal with a
10 signal from a local oscillator to provide an upconverted notched desreading signal;
and mixing means connected to mix the upconverted notched desreading signal
with the spread spectrum signal.

According to another aspect of the present invention, there is provided an
15 apparatus for decoding a spread spectrum signal which has been transmitted via a
transmission channel, said signal comprising a wanted spread spectrum signal
component and an unwanted narrowband signal component, the narrowband signal
having been coded using a coding scheme with error correction capabilities, said
apparatus comprising a narrowband signal subtractor containing

20 means for decoding and correcting the composite signal according to the
narrowband signal coding scheme to provide a decoded version of the narrowband
signal;

means to estimate the amplitude and phase characteristics of the
transmission channel;

25 means to encode the decoded corrected signal according to the
narrowband signal coding scheme to produce an estimated narrowband signal;

means to adjust the phase and amplitude of the estimated narrowband
signal according to the estimated amplitude and phase characteristics; and

→ means to subtract the adjusted signal from the received composite signal
30 to provide an estimate of the spread spectrum signal component.

Preferably the narrowband coding scheme can also detect uncorrectable errors and
in which the decoding means is arranged upon detection of an uncorrectable error
to suppress the said subtraction.

In another embodiment there is also included means for attenuating the frequencies of the signal corresponding to the narrowband signal component wherein the decoding means is arranged upon detection of an uncorrectable error to cause the
5 attenuating means to be used.

Corresponding methods for decoding spread spectrum signals are also provided.

Embodiments of the present invention will be described, by way of example only,
10 with reference to the drawings where

Figure 1 shows the carrier to interference ratios for a high power spectral density narrowband signal and a low power spectral density wideband signal;

Figure 2 shows a block diagram of a known composite signal transmitter using a
15 notch filter to attenuate some frequencies in the transmitted wideband signal;

Figure 3a shows a block diagram of a conventional spread spectrum signal receiver;

Figure 3b shows a block diagram of a known spread spectrum signal receiver with a suppression filter to attenuate some frequencies in the received wideband signal;

20 Figure 4 shows a data signal encoded using a seven bit pseudo noise code;

Figure 5 shows a possible receiver architecture which despreads the signal prior to down conversion;

Figure 6 shows a hypothetical receiver which uses a notch filter to attenuate predetermined frequencies of the received signal prior to downconversion of the
25 signal;

Figure 7 shows a receiver which combines the despreading and filtering of the signal prior to downconversion in accordance with one embodiment of the present invention;

Figure 8 shows combined despreading and filtering means according to an
30 embodiment of the present invention;

Figure 9 shows combined despreading and filtering means according to an embodiment of the present invention;

Figure 9a shows combined despreading filtering and downconversion means according to another embodiment of the present invention;

Figure 10 shows a receiver which has a narrowband signal subtractor according to an embodiment of the present invention;

Figure 11 show details of a narrowband signal subtractor;

Figure 12 shows a more sophisticated narrowband signal subtractor which does
5 not subtract the estimated narrowband signal if uncorrectable errors are detected;
and

Figure 13 shows an embodiment of the present invention in which the estimated narrowband signal is not subtracted if uncorrectable errors are detected, and in which a notch filter is used under those circumstances.

10

Figure 1 shows a narrowband signal transmitter 1 transmitting a narrowband signal 2 and a wideband transmitter 3 transmitting a wideband signal 4. The wideband transmitter and the narrowband transmitter are connected to an adder 5. The wideband signal and the narrowband signal are summed at the adder 5 to create a
15 composite signal 6. Of course, different transmitters in different locations may be used, the effect as far as the present invention is concerned is identical. The composite signal is received by a wideband signal decoder 8 where the signal is decoded producing a new composite signal 9. The spectrum of the original wideband signal 4 is despread by the decoder producing a decoded signal
20 component 4' of the new composite signal 9 and the spectrum of the original narrowband signal 2 is spread to produce a new component 2' of the new composite signal 9. The new composite signal 9 has a carrier to interference ratio 10 for the decoded wideband signal 4'.

25 As far as the narrowband signal is concerned, the composite signal 6 has a carrier to interference ratio 7. To increase the carrier to interference ratio 7 for the narrowband signal 2 a notch filter may be inserted in the wideband signal transmitter 3 to attenuate the frequencies used by the narrowband signal. Such a notch filter will reduce the interference caused to the narrowband signal by the wideband signal.

30 Figure 2 shows a known wideband signal transmitter 15 comprising an encoder 11 connected to a notch filter 12 which is in turn connected to a modulator 16 and upconverter 17. The notch filter 12 is used to filter a wideband signal 13 to produce a notched signal 14 which is modulated and upconverted prior to summing

at the adder 5 with the narrowband signal 2. Again, if different transmitters in different locations are used there is no material effect on the invention.

To increase the carrier to interference ratio 10 for the despread wideband signal 4' 5 a notch filter may be inserted in the wideband signal decoder 8 to attenuate the frequencies used by the narrowband signal. Figure 3a shows a block diagram of a conventional spread spectrum signal receiver which comprises a receiver amplifier 23 which is connected to the wideband signal decoder 8 which comprises a downconverter 18 which is connected to a demodulator 19 which is in turn 10 connected to a despreader 21. Of course, demodulation may take place either before or after despreading. Figure 3b shows another embodiment of a wideband signal decoder 30 in which a notch filter 20 is used before the despreader 21. To minimise the spectrum removed from the wideband signal a notch filter with very sharp cut-offs is required.

15

One method of coding a data signal to generate a spread spectrum signal is to chop up an information bit into smaller increments of time which are known as 'chips'. These chips are generated by imposing a pseudo noise code onto the data signal. Figure 4 shows a digital signal encoded using a seven bit pseudo noise 20 code. In a practical implementation a pseudo noise code would be much longer than this. Pseudo noise codes are selected on the basis of particular mathematical properties which make the encoded signal appear somewhat noiselike. A corresponding decoder which knows which pseudo noise code has been imposed on the data signal is used to despread the spectrum of the spread spectrum signal 25 to retrieve the original data signal.

A disadvantage of despreading the signal at baseband as shown in Figure 3a is that any noise added to the signal by the downconverter 18 or the demodulator 19 will result in interference. It is theoretically possible to despread the received signal 30 before downconversion as shown in Figure 5. In the wideband signal decoder 31 only noise in the narrower bandwidth of the despread signal 9 (see Figure 1) will cause interference, and the result of such interference will be less as the despread signal has a much higher power spectral density than the original wideband signal. However, a suppression filter is required before the despreading step, as shown in

Figure 3b, in order to reduce interference from the narrowband signal 2 (see Figure 1) and it is currently not possible to implement programmable digital notch filters with very sharp cut-offs which operate at the high frequencies required.

- 5 It would also be advantageous to perform the filtering before the downconversion and demodulation because then only components which process the signal before the filter would need the large dynamic range required to avoid noise being introduced due to non-linearities. Figure 6 shows a hypothetical architecture for such a receiver, however, as mentioned previously such an architecture could not
10 be achieved using programmable digital filters in currently available hardware. An arrangement using analogue filters, where the required filter was selected from a bank of available filters would be possible, if unwieldy.

Figure 7 shows a block diagram of a spread spectrum signal receiver in which both
15 the despreading and filtering of the signal are performed together. Such an arrangement has the both the advantages discussed above, namely reduced noise bandwidth and that the only component requiring a large dynamic range is the receiver amplifier 23. The composite means may be implemented as shown in Figure 8. Here a PN code generated by a PN code generator 27 is filtered at
20 baseband by a digital programmable notch filter 26. The filtered PN code is then used to despread the incoming signal using a mixer 25, thus achieving the despreading and filtering of the incoming signal simultaneously, at the received frequency. After mixing by the mixer 25 the signal is filtered by a band-pass filter 35 and amplified by an amplifier 36.

25

Figure 9 shows an improved arrangement which can be implemented with currently available hardware. Mixers which are optimised to operate with one high frequency and one low frequency input, such as is required for the mixer of Figure 8, usually require a high power high frequency signal. However, in the embodiment shown in
30 Figure 8 the high frequency signal is a low power signal from an antenna. In the embodiment shown in Figure 9, the mixer 25 is replaced by two mixers 28 and 29 and a local oscillator 34. The mixer 28 therefore receives a high power signal from the local oscillator 34 which is mixed with the filtered PN code received from the notch filter 26. This high power signal is then mixed with the low power signal

received from the antenna amplifier 23. In this arrangement the received signal is simultaneously filtered downconverted and despread, so the downconverter 18 may be omitted or if further downconversion is required the downconverter 18 may be retained. The mixer 28 and the mixer 29 may be replaced with a single
5 mixer.

In the arrangement shown in Figure 9 the signal produced by the mixer 28 is symmetrical about a centre frequency (F_c) and has two ranges of attenuated frequencies at $F_c \pm$ the frequencies attenuated by the attenuating notch filter 26.
10 The result of this is that twice the amount of the spectrum is removed than is necessary. If a further upconversion stage 50,51 is employed as shown in Figure 9a then the PN code is upconverted to an intermediate frequency within the operating range of the notch filter 28 so that the signal produced by the mixer 28 only has one range of attenuated frequencies.

15

Figure 10 shows a block diagram for another receiver 39 for decoding a composite signal comprising a high power narrowband signal component which has been coded using a coding scheme with some error correction capabilities; and a low power spread spectrum signal component. The narrowband signal subtractor 40
20 removes the high power narrowband signal component prior to despreading the spread spectrum signal component. In the narrowband signal subtractor 40, the composite signal is decoded according to the narrowband signal coding scheme, any errors introduced into the narrowband signal as a result of the low level spread spectrum signal are corrected at this stage and the resultant narrowband signal
25 estimate is subtracted from the composite signal, resulting in a wideband signal which is a good estimate of the transmitted spread spectrum signal. Clearly the receiver 39 would be advantageous for receiving a signal in which the frequencies in the narrowband frequency range were not attenuated by the transmitter according to figure 2.

30

It would be possible to perform the narrowband signal subtraction after downconversion and demodulation. However, if the narrowband signal subtraction is done prior to downconversion and demodulation then there is the advantage, as

discussed previously, that only the receiver amplifier 23 need to have a large dynamic range.

Figure 11 shows an embodiment of the narrowband signal subtractor 40 where the narrowband signal is coded using GSM. A received composite signal is demodulated by a GSM demodulator 42. The GSM demodulator 42 includes forward error correction and deinterleaving, as required. During demodulation error correction occurs, thus removing some or all of the errors introduced by any spread spectrum signal component. The demodulated corrected signal is then modulated by a GSM modulator 44 to produce an estimate of the transmitted GSM signal. The GSM modulator provides forward error correction bits and interleaving as necessary. In parallel with the modulation and demodulation steps the received composite signal is processed by a GSM channel estimator 43. This provides an estimate of the phase and amplitude changes and multipath delay introduced by the transmission channel. These estimates are used to adjust the phase and amplitude of the estimated signal by an inverse equaliser 45. This adjusted signal is subtracted by a subtractor 46 from a delayed version of the received composite signal. A delay element 41 compensates for the delay introduced by the elements 42, 44, 45 and 43. The GSM demodulator 42, the GSM modulator 44 and the GSM channel estimator 43 may be implemented using commonly available standard GSM microchips. The inverse equaliser 45 may be implemented using a programmable digital filter.

In practice it may be necessary to perform the inverse equalisation at a lower frequency than the received frequency, in which case the upconversion which is included in the GSM modulator 44 would be carried out instead after the inverse equalisation performed at 45.

Figure 12 show a block diagram of an improved narrowband signal subtractor 40' which can be used if the narrowband signal coding scheme can detect uncorrectable errors. If an uncorrectable error is detected by the demodulator 42 then an output 50 is used to control a switch 51 and the narrowband signal subtraction at 46 does not occur.

Figure 13 shows a block diagram of another embodiment of a narrowband signal subtractor 40'' connected to a combined despreader and notch filter 24'' which replaces the despreader 21 (Figure 10) an improved receiver. In this embodiment the output 50 is used to control both the switch 51 to bypass narrowband signal subtraction and to control a switch 52 to control whether the notch filter 26 is bypassed. When an uncorrectable error is detected the narrowband signal subtraction is bypassed and the notch filter 26 is used instead.

↓
or narrowband

↓
to clear
narrowband error

CLAIMS

1. An apparatus for decoding a spread spectrum signal comprising
means for generating a despreading signal for despreading the
5 spectrum of a signal;
means for despreading the spectrum of a signal using the despreading
signal;
means for downconverting a signal to produce a lower frequency signal;
and
10 means for attenuating predetermined frequencies of a signal;
wherein the despreading means is connected to despread the spread spectrum
signal prior to downconverting and the attenuating means is connected to
attenuate predetermined frequencies of the spread spectrum signal prior to
downconverting.
- 15
2. An apparatus according to claim 1 in which the received signal is a
composite signal comprising a spread spectrum signal component and a
narrowband signal component which has a much higher power spectral density
than the spread spectrum signal component and the attenuated frequencies are
20 those present in the narrowband signal component.
3. An apparatus according to claim 1 or claim 2 in which the spectrum is de-
spread using a pseudo noise code.
- 25 4. An apparatus according to any of the preceding claims in which the
attenuating means is connected to attenuate predetermined frequencies of the
despreading signal and the resulting notched despreading signal is connected to
simultaneously despread the spread spectrum signal and attenuate the
predetermined frequencies of the spread spectrum signal.

5. An apparatus (33) for decoding a spread spectrum signal comprising means (27) for generating a despread signal for desreading the spectrum of the signal; and
attenuating means (26) connected to attenuate predetermined frequencies
5 of the desreading signal to provide a notched desreading signal;
characterised in that the apparatus further comprises
mixing means (28) connected to mix the notched desreading signal with a signal from a local oscillator (34) to provide an upconverted notched desreading signal; and
10 mixing means (29) connected to mix the upconverted notched desreading signal with the spread spectrum signal.
6. An apparatus for decoding a spread spectrum signal which has been
15 transmitted via a transmission channel, said signal comprising a wanted spread spectrum signal component and an unwanted narrowband signal component, the narrowband signal having been coded using a coding scheme with error correction capabilities, said apparatus comprising a narrowband signal subtractor containing
means for decoding and correcting the composite signal according to the
20 narrowband signal coding scheme to provide a decoded version of the narrowband signal;
means to estimate the amplitude and phase characteristics of the transmission channel;
means to encode the decoded corrected signal according to the
25 narrowband signal coding scheme to produce an estimated narrowband signal;
means to adjust the phase and amplitude of the estimated narrowband signal according to the estimated amplitude and phase characteristics; and
means to subtract the adjusted signal from the received composite signal to provide an estimate of the spread spectrum signal component.
- 30
7. An apparatus according to claim 6 wherein the narrowband coding scheme can also detect uncorrectable errors and in which the decoding means is arranged upon detection of an uncorrectable error to suppress the said subtraction.

8. An apparatus according to claim 7 further comprising means for attenuating the frequencies of the signal corresponding to the narrowband signal component wherein the decoding means is arranged upon detection of an uncorrectable error to cause the attenuating means to be used.

5

9. A method for decoding a spread spectrum signal comprising the steps of generating a despreading signal for despreading the spectrum of a signal; despreading the spectrum of a signal using the despreading signal; downconverting a signal to produce a lower frequency signal; and

10

attenuating predetermined frequencies of a signal;

wherein despreading of the spread spectrum signal is performed prior to downconverting and attenuating of predetermined frequencies of the spread spectrum signal is performed prior to downconverting.

15

10. A method according to claim 9 in which the received signal is a composite signal comprising a spread spectrum signal component and a narrowband signal component which has a much higher power spectral density than the spread spectrum signal component and the attenuated frequencies are those present in the narrowband signal component.

20

11. A method according to claim 9 or claim 10 in which the spectrum is despread using a pseudo noise code.

25

12. A method according to any one of claim 9 to claim 11 in which the predetermined frequencies of the despreading signal are attenuated and the resulting notched despreading signal is used to simultaneously despread the spread spectrum signal and attenuate the predetermined frequencies of the spread spectrum signal.

30

13. A method for decoding a spread spectrum signal comprising the steps of generating a despreading signal for despreading the spectrum of the signal; and

attenuating predetermined frequencies of the despreading signal to provide a notched despreading signal;

characterised in that the method further comprises steps of
mixing the notched despread signal with a signal from a local oscillator
to provide an upconverted notched despread signal; and
mixing the upconverted notched despread signal with the spread
5 spectrum signal.

14. A method for decoding a spread spectrum signal which has been
transmitted via a transmission channel, said signal comprising a wanted spread
spectrum signal component and an unwanted narrowband signal component, the
10 narrowband signal having been coded using a coding scheme with error correction
capabilities, said method including narrowband signal subtraction comprising the
steps of

decoding and correcting the composite signal according to the narrowband
signal coding scheme to provide a decoded version of the narrowband signal;
15 estimating the amplitude and phase characteristics of the transmission
channel;

encoding the decoded corrected signal according to the narrowband signal
coding scheme to produce an estimated narrowband signal;

adjusting the phase and amplitude of the estimated narrowband signal
20 according to the estimated amplitude and phase characteristics; and

subtracting the adjusted signal from the received composite signal to
provide an estimate of the spread spectrum signal component.

15. A method according to claim 14 wherein the narrowband coding scheme
25 can also detect uncorrectable errors and in which upon detection of an
uncorrectable error the step of subtracting the narrowband signal is omitted.

16. A method according to claim 15 including the optional step of attenuating
the frequencies of the signal corresponding to the narrowband signal component
30 wherein upon detection of an uncorrectable error the attenuation step is
performed.

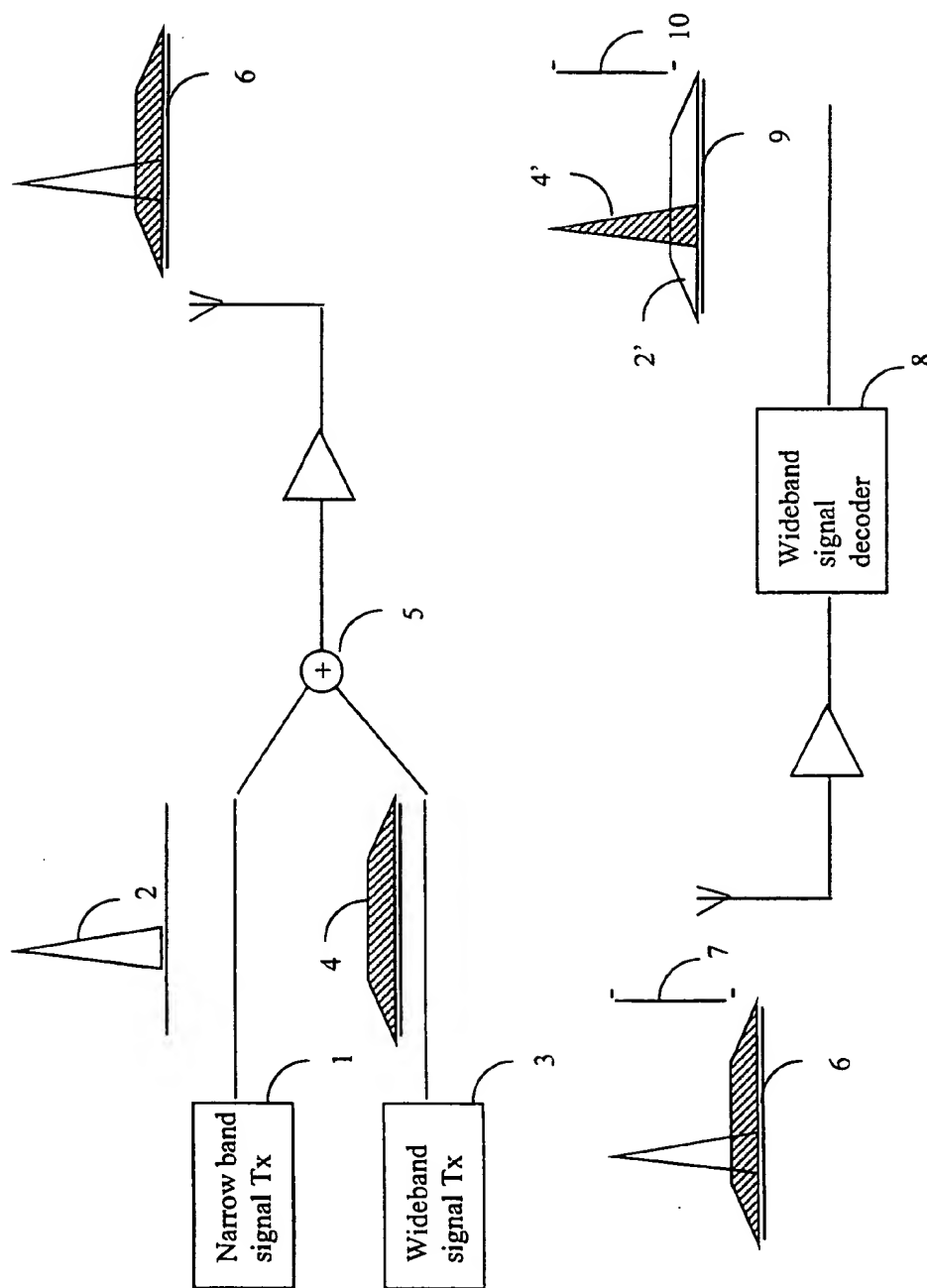


Figure 1

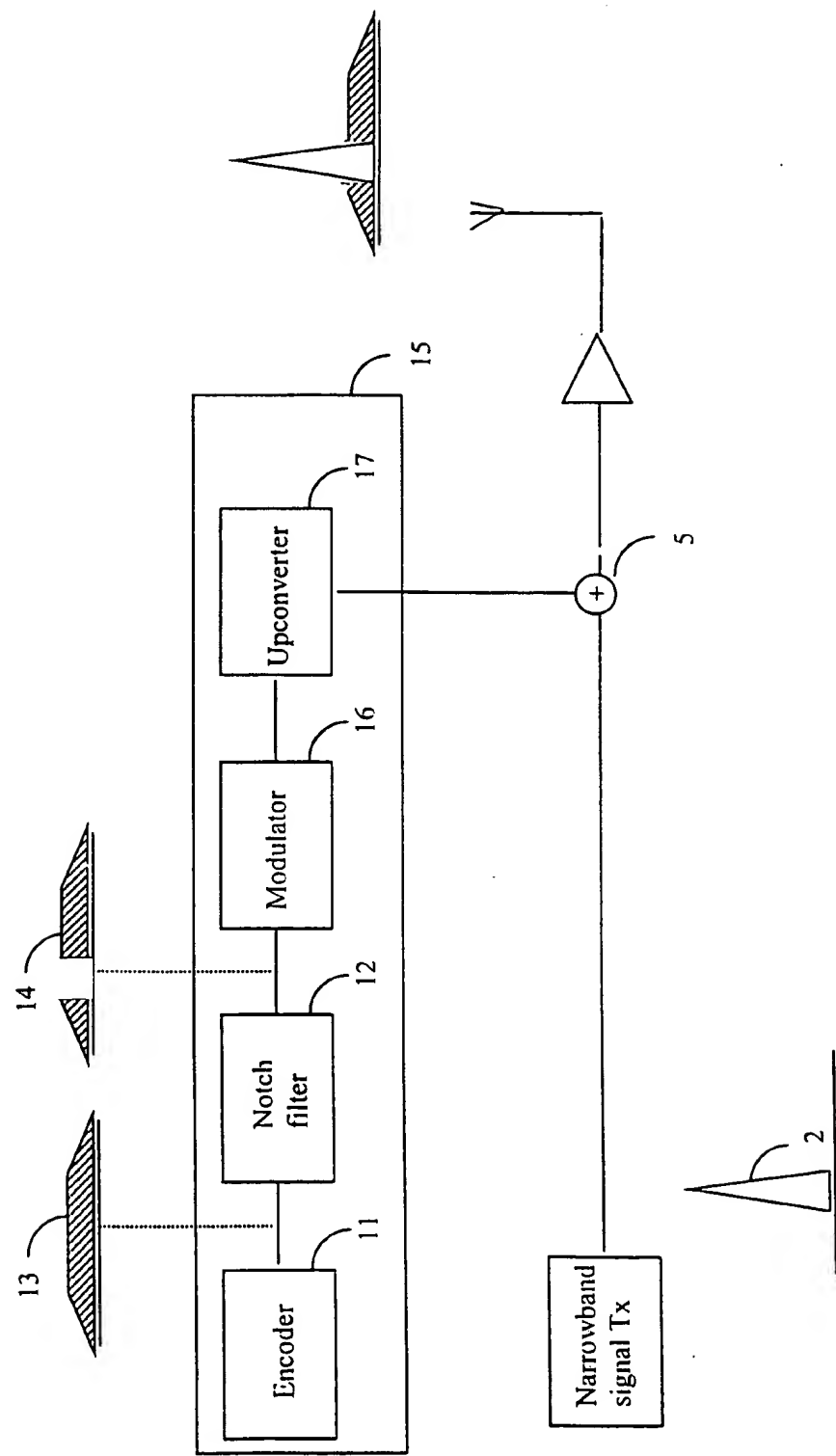


Figure 2

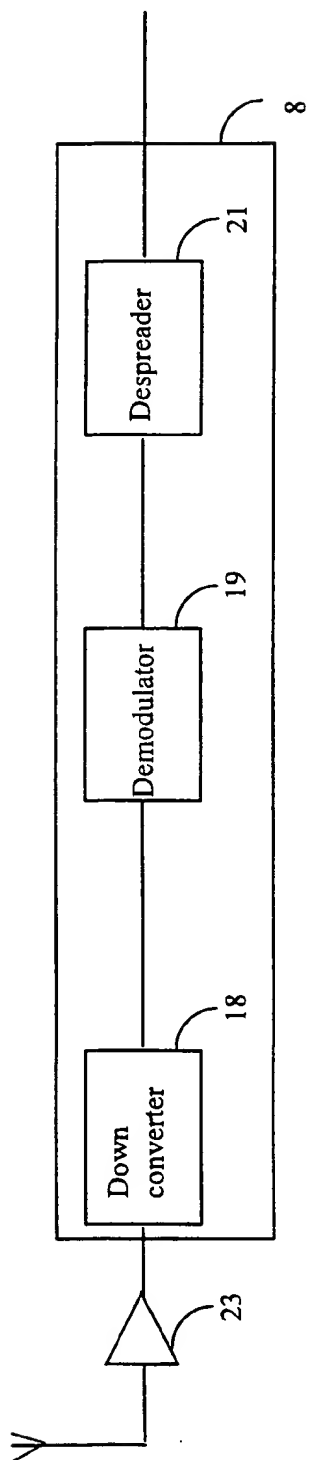


Figure 3a

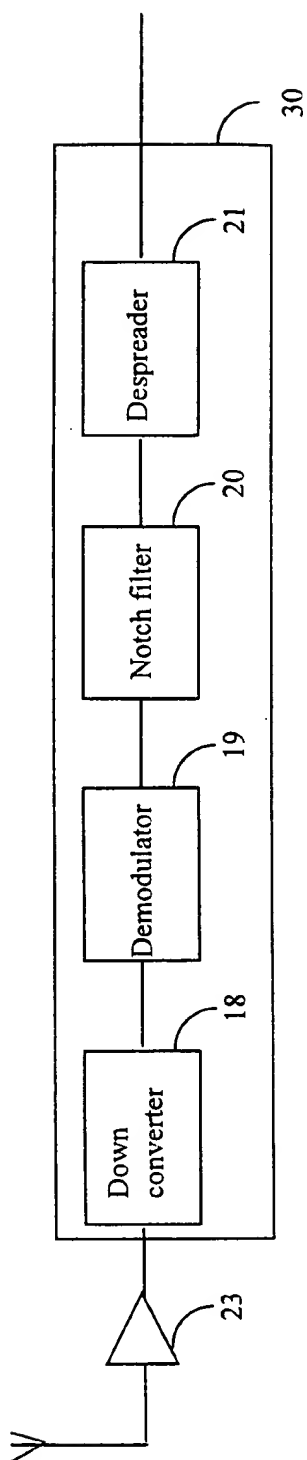


Figure 3b

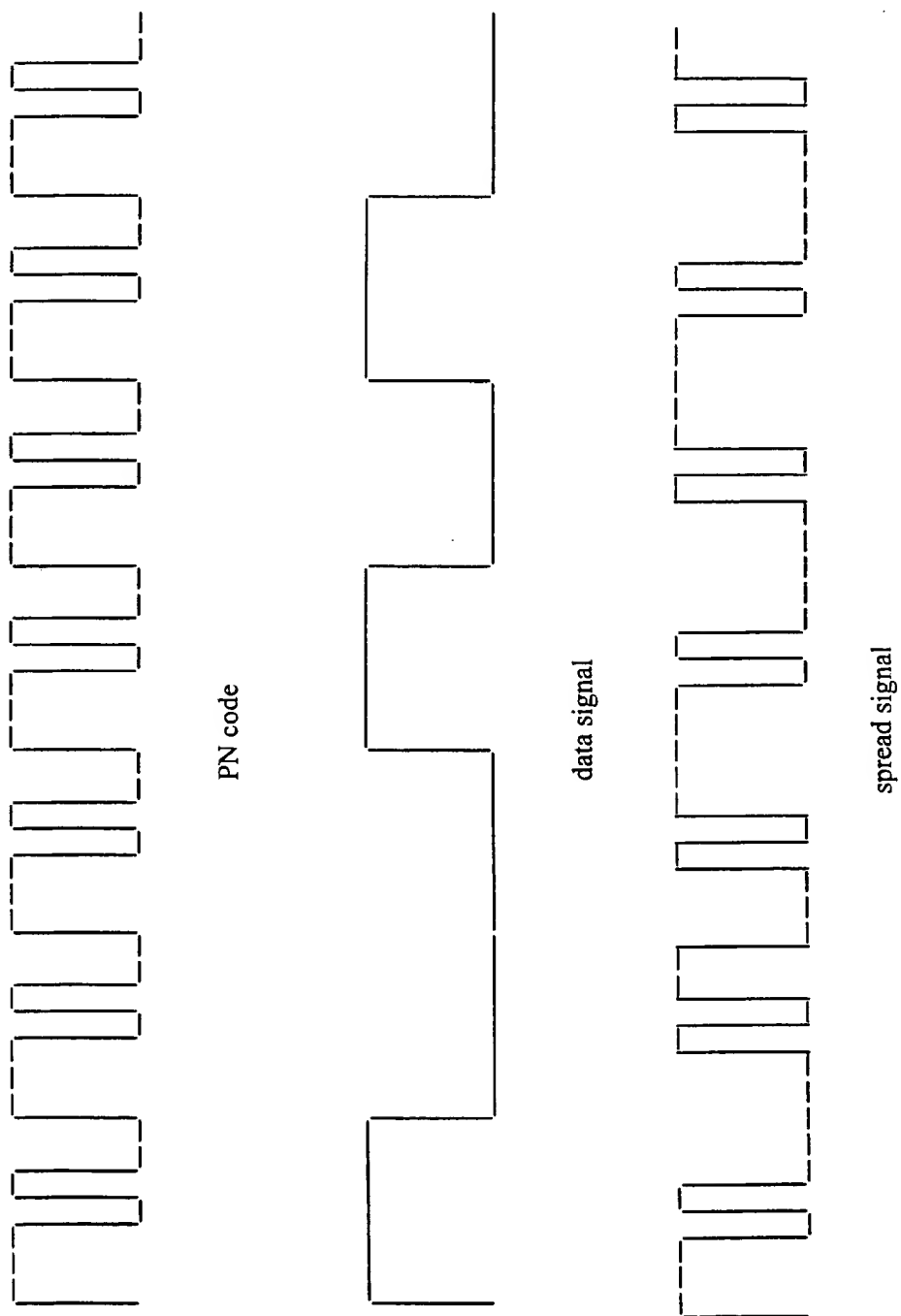


Figure 4

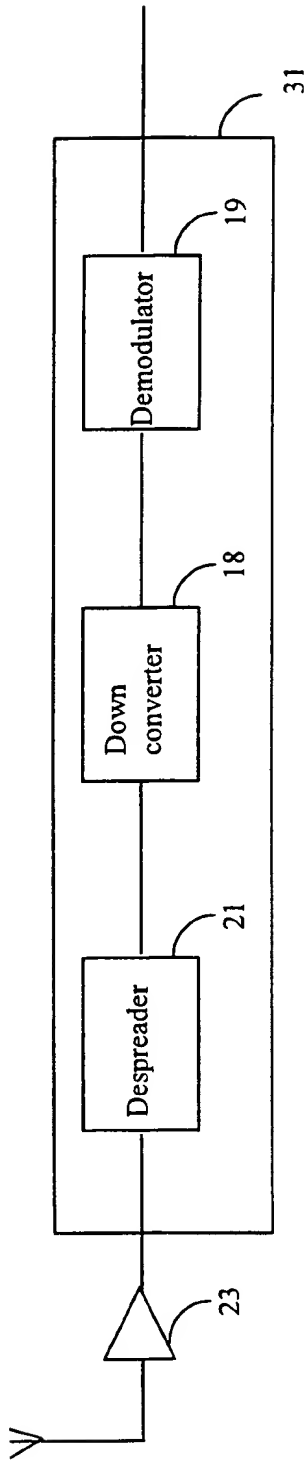


Figure 5

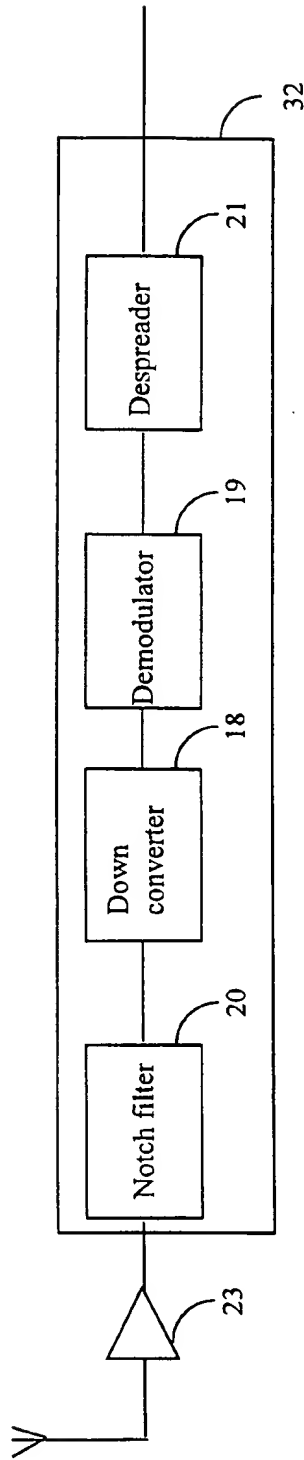


Figure 6

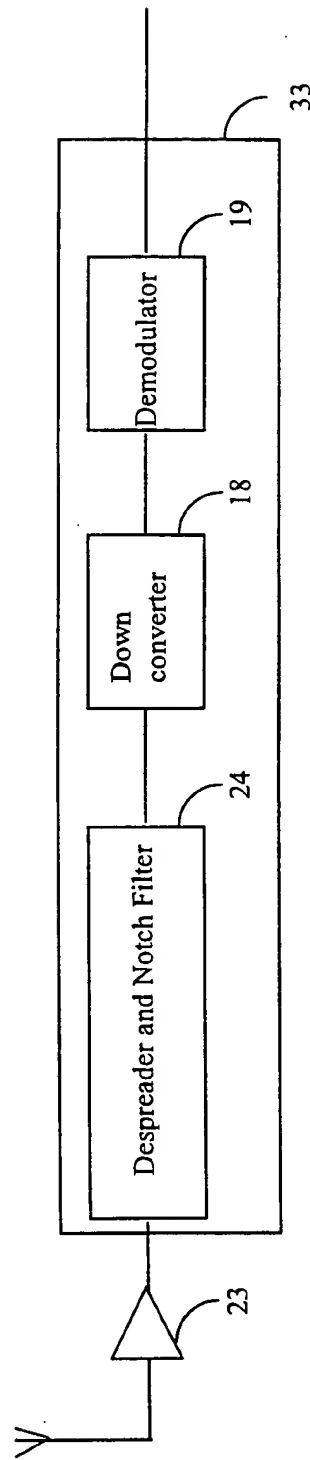


Figure 7

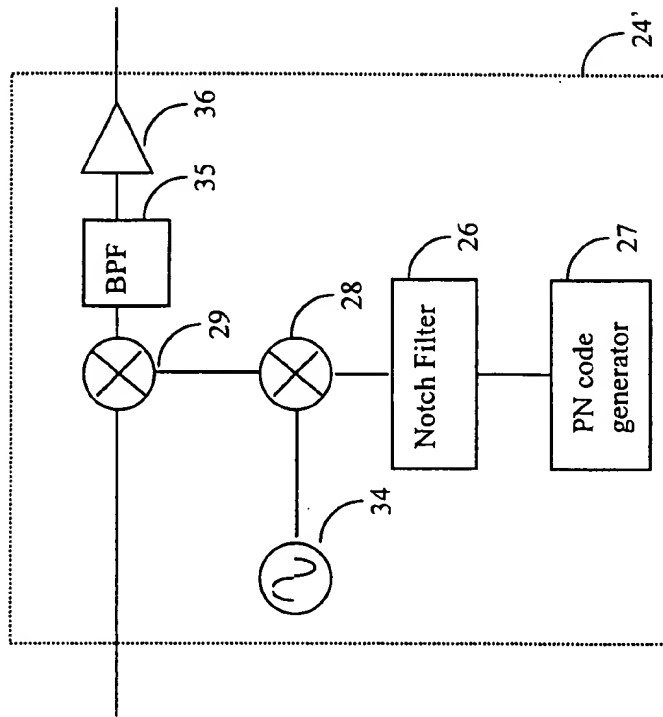


Figure 9

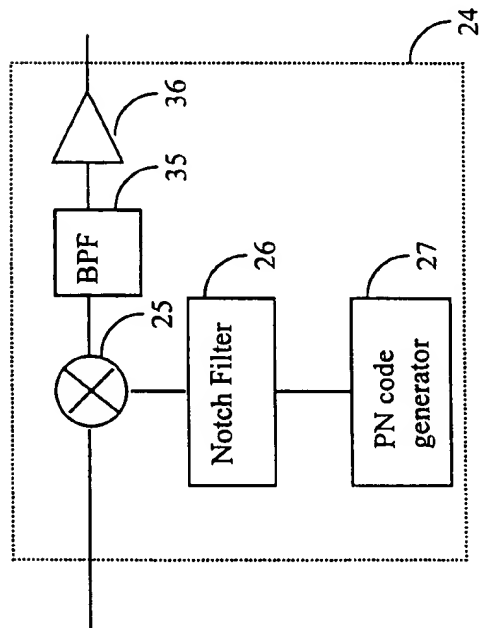


Figure 8

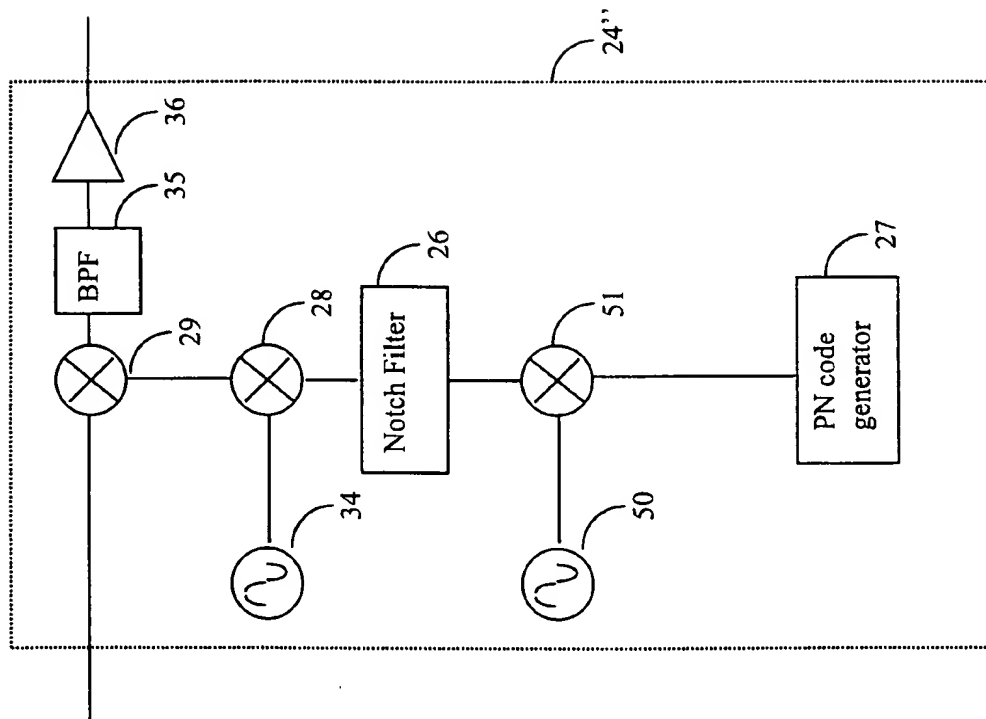


Figure 9a

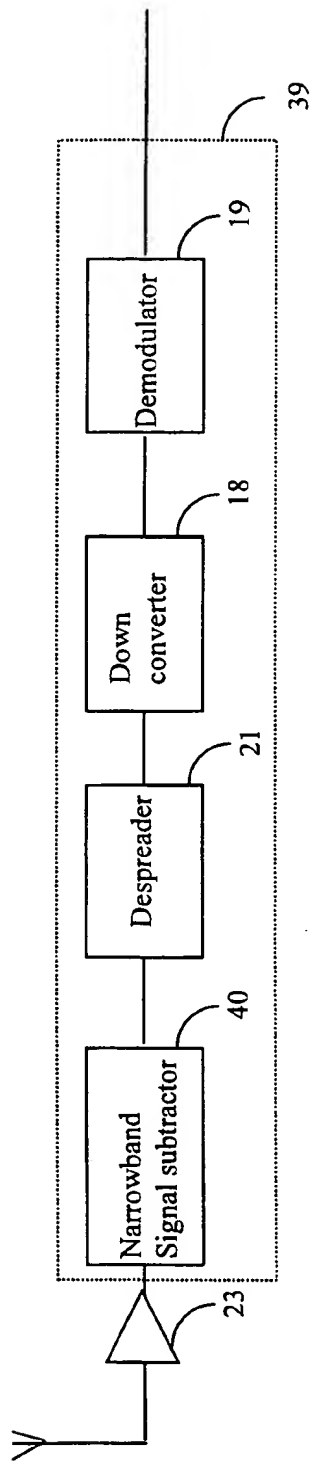


Figure 10

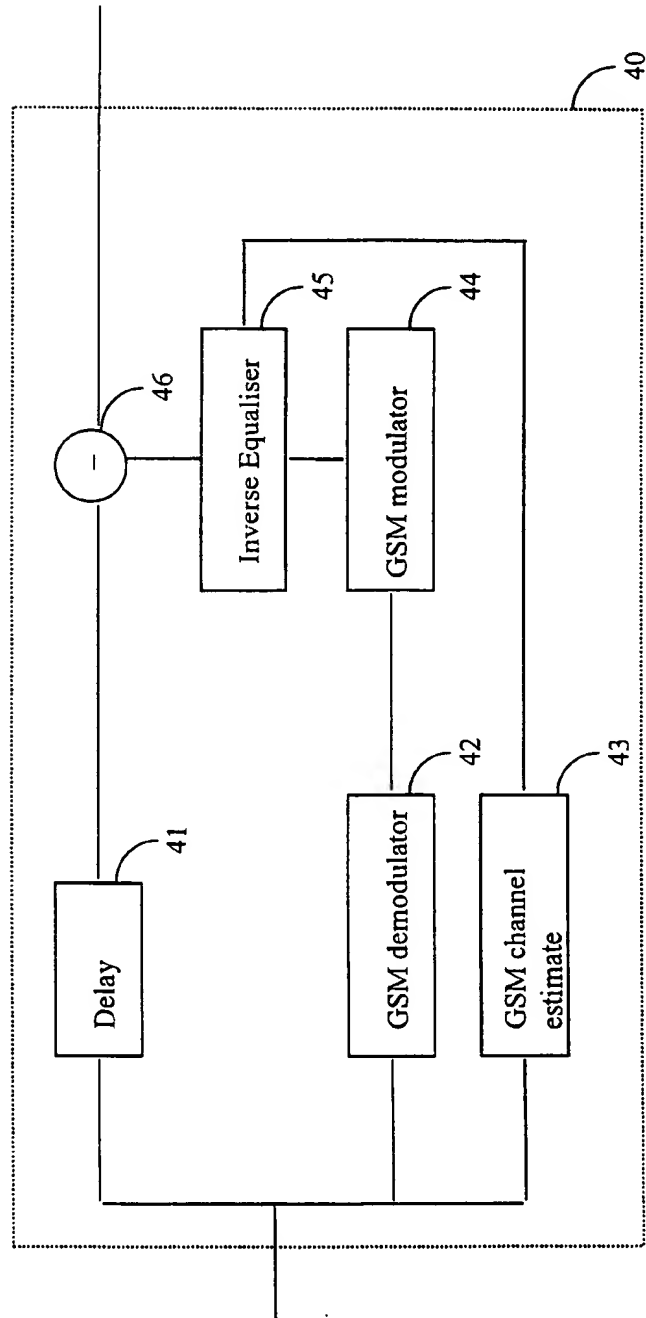


Figure 11

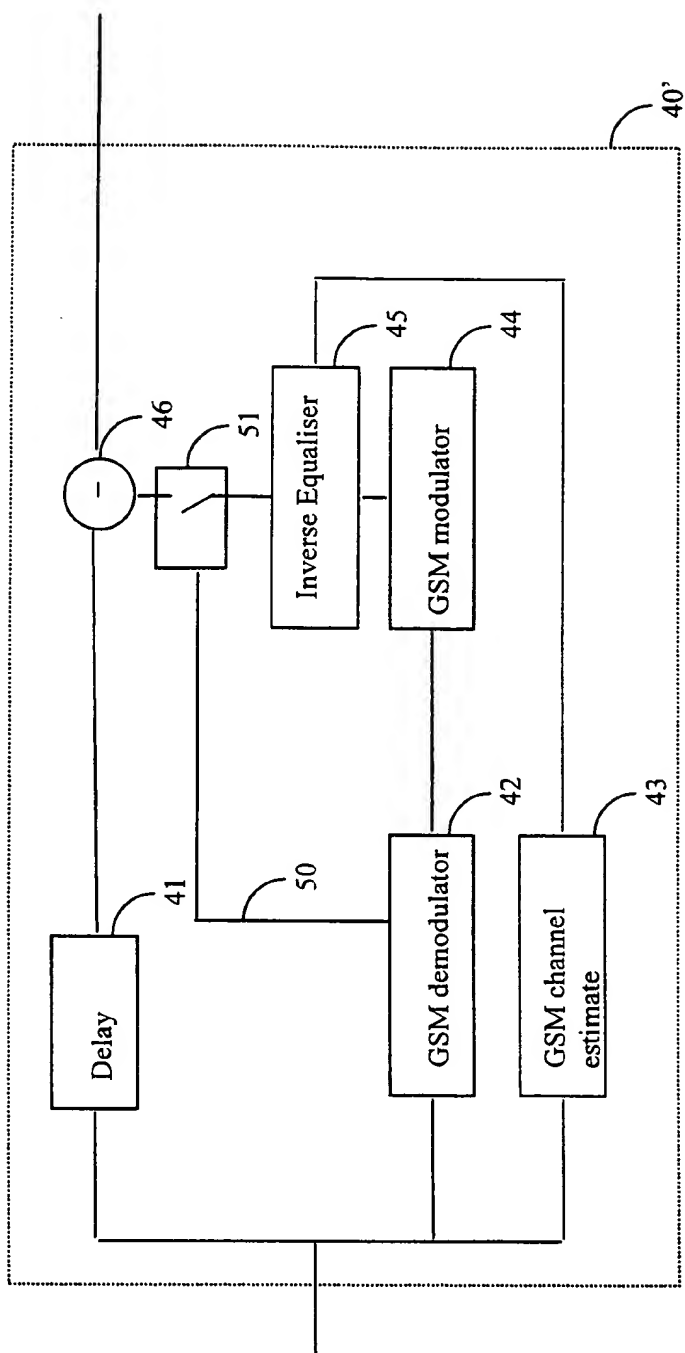


Figure 12

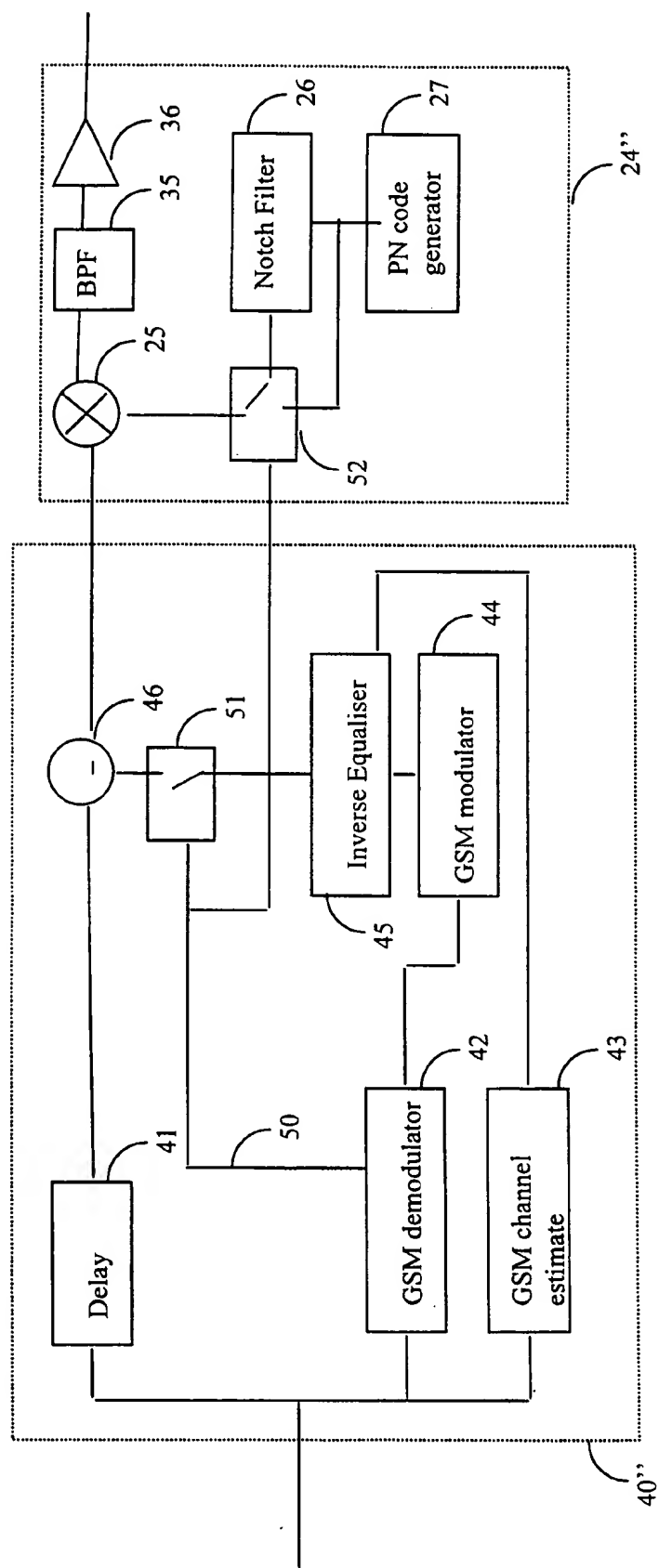


Figure 13